



EUROPEAN
TELECOMMUNICATION
STANDARD

ETS 300 575

September 1994

Source: ETSI TC-SMG

Reference: GSM 05.03

ICS: 33.060.30

Key words: European digital cellular telecommunications system, Global System for Mobile communications (GSM)

**European digital cellular telecommunications system (Phase 2);
Channel coding
(GSM 05.03)**

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Foreword

This European Telecommunication Standard (ETS) has been produced by the Special Mobile Group (SMG) Technical Committee (TC) of the European Telecommunications Standards Institute (ETSI).

This ETS specifies the channel coding of used within the European digital cellular telecommunications system (Phase 2).

This ETS correspond to GSM technical specification, GSM 05.03 version 4.1.3.

The specification from which this ETS has been derived was originally based on CEPT documentation, hence the presentation of this ETS may not be entirely in accordance with the ETSI/PNE rules.

Reference is made within this ETS to GSM Technical Specifications (GSM-TSs) (NOTE).

NOTE: TC-SMG has produced documents which give the technical specifications for the implementation of the European digital cellular telecommunications system. Historically, these documents have been identified as GSM Technical Specifications (GSM-TSs). These TSs may have subsequently become I-ETTs (Phase 1), or ETs (Phase 2), whilst others may become ETSI Technical Reports (ETRs). GSM-TSs are, for editorial reasons, still referred to in GSM ETs.

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1.1 Scope

A reference configuration of the transmission chain is shown in GSM 05.01. According to this reference configuration, this technical specification specifies the data blocks given to the encryption unit.

It includes the specification of encoding, reordering, interleaving and the stealing flag. It does not specify the channel decoding method.

The definition is given for each kind of logical channel, starting from the data provided to the channel encoder by the speech coder, the data terminal equipment, or the controller of the MS or BS. The definitions of the logical channel types used in this technical specification are given in GSM 05.02, a summary is in Annex A.

1.2 Normative references

This ETS incorporates by dated and undated reference, provisions from other publications. These normative references are cited at the appropriate places in the text and the publications are listed hereafter. For dated references, subsequent amendments to or revisions of any of these publications apply to this ETS only when incorporated in it by amendment or revision. For undated references, the latest edition of the publication referred to applies.

- [1] GSM 01.04 (ETR 100): "European digital cellular telecommunication system (Phase 2); Definitions, abbreviations and acronyms".
- [2] GSM 04.08 (ETS 300 557): "European digital cellular telecommunication system (Phase 2); Mobile radio interface layer 3 specification".
- [3] GSM 04.21 (ETS 300 562): "European digital cellular telecommunication system (Phase 2); Rate adaption on the Mobile Station - Base Station System (MS - BSS) interface".
- [4] GSM 05.01 (ETS 300 573): "European digital cellular telecommunication system (Phase 2); Physical layer on the radio path General description".
- [5] GSM 05.02 (ETS 300 574): "European digital cellular telecommunication system (Phase 2); Multiplexing and multiple access on the radio path".
- [6] GSM 06.10 (ETS 300 580-2): "European digital cellular telecommunication system (Phase 2); Full rate speech transcoding".

1.3 Definitions and abbreviations

Definitions and abbreviations used in this specification are listed in GSM 01.04.

2. General

2.1 General Organization

Each channel has its own coding and interleaving scheme. However, the channel coding and interleaving is organized in such a way as to allow, as much as possible, a unified decoder structure.

Each channel uses the following sequence and order of operations:

- The information bits are coded with a systematic block code, building words of information + parity bits.
- These information + parity bits are encoded with a convolutional code, building the coded bits.
- Reordering and interleaving the coded bits, and adding a stealing flag, gives the interleaved bits.

All these operations are made block by block, the size of which depends on the channel. However, most of the channels use a block of 456 coded bits which is interleaved and mapped onto bursts in a very similar way for all of them. Figure 1 gives a diagram showing the general structure of the channel coding.

This block of 456 coded bits is the basic structure of the channel coding scheme. In the case of speech TCH, this block carries the information of one speech frame. In case of control channels, it carries one message.

In the case of Fast ACCH, a coded message block of 456 bits is divided into eight sub-blocks. The first four sub-blocks are sent by stealing the even numbered bits of four timeslots in consecutive frames used for the TCH. The other four sub-blocks are sent by stealing the odd numbered bits of the relevant timeslot in four consecutive used frames delayed 2 or 4 frames relative to the first frame. Along with each block of 456 coded bits there is, in addition, a stealing flag (8 bits), indicating whether the block belongs to the TCH or to the fast ACCH. In the case of slow ACCH, BCCH or CCCH, this stealing flag is dummy.

Some cases do not fit in the general organization, and do not use the block of 456 coded bits. They are the random access messages of the RACH on uplink and the synchronization information broadcast on the SCH.

2.2 Naming Convention

For ease of understanding a naming convention for bits is given for use throughout the technical specification:

- General naming

"k" and "j" for numbering of bits in data blocks and bursts.

"K_x" gives the amount of bits in one block, where "x" refers to the data type

"n" is used for numbering of delivered data blocks where

"N" marks a certain data block

"B" is used for numbering of bursts or blocks where

"B₀" marks the first burst or block carrying bits from the data block with n = 0 (first data block in the transmission)

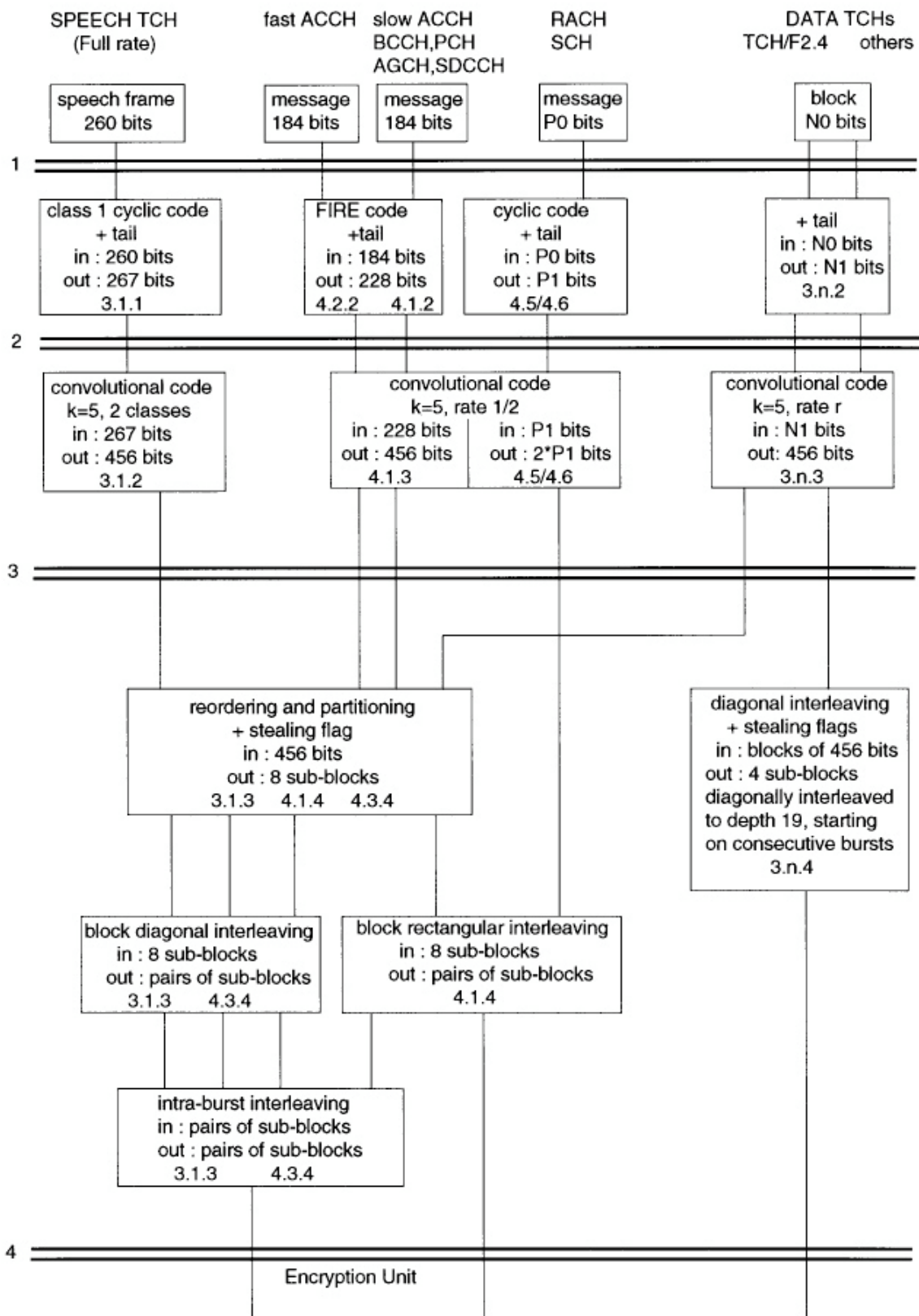


Figure 1 : Channel Coding and Interleaving Organisation

In the case of DATA TCHs. N0, N1, n depends on the type of data TCH.
 In each box, the last line indicates the chapter defining the function.

- Interface 1: Information bits (d)
- 2: Information + parity bits (u)
- 3: coded bits (c)
- 4: interleaved bits (e)

- Data delivered to the encoding unit (interface 1 in figure 1):

$$d(n, k) \text{ or } d(k) \quad \text{for} \quad \begin{array}{l} k = 0, 1, \dots, K_d - 1 \\ n = 0, 1, \dots, N, N+1, \dots \end{array}$$

- Data after the first encoding step (block code, cyclic code; interface 2 in figure 1):

$$u(n, k) \text{ or } u(k) \quad \text{for} \quad \begin{array}{l} k = 0, 1, \dots, K_u - 1 \\ n = 0, 1, \dots, N, N+1, \dots \end{array}$$

- Data after the second encoding step (convolutional code ; interface 3 in figure 1):

$$c(n, k) \text{ or } c(k) \quad \text{for} \quad \begin{array}{l} k = 0, 1, \dots, K_c - 1 \\ n = 0, 1, \dots, N, N+1, \dots \end{array}$$

- Interleaved data:

$$i(B, k) \text{ for} \quad \begin{array}{l} k = 0, 1, \dots, K_i - 1 \\ B = B_0, B_0 + 1, \dots \end{array}$$

- Bits in one burst (interface 4 in figure 1):

$$e(B, k) \text{ for} \quad \begin{array}{l} k = 0, 1, \dots, 114, 115 \\ B = B_0, B_0 + 1, \dots \end{array}$$

3 Traffic Channels (TCH)

Two kinds of traffic channel are considered: speech and data. Both of them use the same general structure (see fig.1), and in both cases, a piece of information can be stolen by the fast ACCH.

3.1 Speech channel at full rate (TCH/FS)

The speech coder delivers to the channel encoder a sequence of blocks of data. In case of a full rate speech TCH, one block of data corresponds to one speech frame. Each block contains 260 information bits, including 182 bits of class 1 (protected bits), and 78 bits of class 2 (no protection), (see Table 2).

The bits delivered by the speech coder are received in the order indicated in GSM 06.10 and have to be rearranged according to Table 2 before channel coding as defined in 3.1.1 to 3.1.4. The rearranged bits are labelled $\{d(0), d(1), \dots, d(259)\}$, defined in the order of decreasing importance.

3.1.1 Parity and tailing for a speech frame

- a) Parity bits:

The first 50 bits of class 1 are protected by three parity bits used for error detection. These parity bits are added to the 50 bits, according to a degenerate (shortened) cyclic code (53,50,2), using the generator polynomial:

$$g(D) = D^3 + D + 1$$

The encoding of the cyclic code is performed in a systematic form, which means that, in $GF(2)$, the polynomial:

$$d(0)D^{52} + d(1)D^{51} + \dots + d(49)D^3 + p(0)D^2 + p(1)D + p(2)$$

where $p(0)$, $p(1)$, $p(2)$ are the parity bits, when divided by $g(D)$, yields a remainder equal to $1 + D + D^2$

b) Tailing bits and reordering:

The information and parity bits of class 1 are reordered, defining 189 information + parity + tail bits of class 1, $\{u(0), u(1), \dots, u(188)\}$ defined by:

$$\begin{aligned} u(k) &= d(2k) \quad \text{and} \quad u(184-k) = d(2k+1) && \text{for } k = 0, 1, \dots, 90 \\ u(91+k) &= p(k) && \text{for } k = 0, 1, 2 \\ u(k) &= 0 \text{ (tail bits)} && \text{for } k = 185, 186, 187, 188 \end{aligned}$$

3.1.2 Convolutional encoder

The class 1 bits are encoded with the 1/2 rate convolutional code defined by the polynomials:

$$\begin{aligned} G_0 &= 1 + D^3 + D^4 \\ G_1 &= 1 + D + D^3 + D^4 \end{aligned}$$

The coded bits $\{c(0), c(1), \dots, c(455)\}$ are then defined by:

$$\begin{aligned} \text{- class 1 : } \quad c(2k) &= u(k) + u(k-3) + u(k-4) \\ c(2k+1) &= u(k) + u(k-1) + u(k-3) + u(k-4) && \text{for } k = 0, 1, \dots, 188 \\ &&& u(k) = 0, k < 0 \\ \text{- class 2 : } \quad c(378+k) &= d(182+k) && \text{for } k = 0, 1, \dots, 77 \end{aligned}$$

3.1.3 Interleaving

The coded bits are reordered and interleaved according to the following rule :

$$\begin{aligned} i(B, j) &= c(n, k), && \text{for } k = 0, 1, \dots, 455 \\ &&& n = 0, 1, \dots, N, N+1, \dots \\ &&& B = B_0 + 4n + k \bmod (8) \\ &&& j = 2[(49k) \bmod 57] + [(k \bmod 8) \text{ div } 4] \end{aligned}$$

The result of the interleaving is a distribution of the reordered 456 bits of a given data block, $n = N$, over 8 blocks using the even numbered bits of the first 4 blocks ($B = B_0 + 4N + 0, 1, 2, 3$) and odd numbered bits of the last 4 blocks ($B = B_0 + 4N + 4, 5, 6, 7$). The reordered bits of the following data block, $n = N+1$, use the even numbered bits of the blocks $B = B_0 + 4N + 4, 5, 6, 7$ ($B = B_0 + 4(N+1) + 0, 1, 2, 3$) and the odd numbered bits of the blocks $B = B_0 + 4(N+1) + 4, 5, 6, 7$. Continuing with the next data blocks shows that one block always carries 57 bits of data from one data block ($n = N$) and 57 bits of data from the next block ($n = N+1$), where the bits from the data block with the higher number always are the even numbered data bits, and those of the data block with the lower number are the odd numbered bits.

The block of coded data is interleaved "block diagonal", where a new data block starts every 4th block and is distributed over 8 blocks.

3.1.4 Mapping on a Burst

The mapping is given by the rule :

$$\begin{aligned} e(B, j) &= i(B, j) && \text{and} && e(B, 59+j) = i(B, 57 + j) && \text{for } j = 0, \dots, 56 \\ \text{and} &&& && && \\ e(B, 57) &= hl(B) && \text{and} && e(B, 58) = hu(B) \end{aligned}$$

The two bits, labelled $hl(B)$ and $hu(B)$ on burst number B are flags used for indication of control channel signalling. For each TCH/FS block not stolen for signalling purposes:

$$\begin{aligned} hu(B) &= 0 && \text{for the first 4 burst (indicating status of even numbered bits)} \\ hl(B) &= 0 && \text{for the last 4 bursts (indicating status of odd numbered bits)} \end{aligned}$$

For the use of $hl(B)$ and $hu(B)$ when a speech frame is stolen for signalling purposes see section 4.2.5.

3.2 Speech channel at half rate (TCH/HS)

To be defined for a future evolution of the system.

3.3 Data channel at full rate, 12.0 kbit/s radio interface rate (9.6 kbit/s services (TCH/F9.6))

The definition of a 12.0 kbit/s radio interface rate data flow for data services is given in GSM 04.21.

3.3.1 Interface with user unit

The user unit delivers to the encoder a bit stream organized in blocks of 60 information bits (data frames) every 5 ms. Four such blocks are dealt with together in the coding process $\{d(0), \dots, d(239)\}$. For non-transparent services those four blocks will align with one 240-bit RLP frame.

3.3.2 Block code

The block of $4 * 60$ information bits is not encoded, but only increased with 4 tail bits equal to 0 at the end of the block.

$$\begin{aligned} u(k) &= d(k) & k &= 0, \dots, 239 \\ u(k) &= 0 & k &= 240, \dots, 243 \end{aligned}$$

3.3.3 Convolutional encoder

This block of 244 bits $\{u(0), \dots, u(243)\}$ is encoded with the 1/2 rate convolutional code defined by the following polynomials:

$$\begin{aligned} G_0 &= 1 + D^3 + D^4 \\ G_1 &= 1 + D + D^3 + D^4 \end{aligned}$$

resulting in 488 coded bits $\{C(0), C(1), \dots, C(487)\}$ with

$$\begin{aligned} C(2k) &= u(k) + u(k-3) + u(k-4) \\ C(2k+1) &= u(k) + u(k-1) + u(k-3) + u(k-4) \quad \text{for } k = 0, \dots, 243; u(k) = 0, k < 0 \end{aligned}$$

The code is punctured in such a way that the following 32 coded bits:

$$\{C(11 + 15j); j = 0, \dots, 31\} \text{ are not transmitted.}$$

The result is a block of 456 coded bits, $\{c(0), \dots, c(455)\}$

3.3.4 Interleaving

The coded bits are reordered and interleaved according to the following rule :

$$\begin{aligned} i(B, j) &= c(n, k) & \text{for } & k = 0, \dots, 455 \\ & & & n = 0, 1, \dots, N, N+1, \dots \\ & & & B = B_0 + 4n + k \bmod (19) + k \operatorname{div} 114 \\ & & & j = k \bmod (19) + 19 [k \bmod (6)] \end{aligned}$$

The result of the interleaving is a distribution of the reordered 114 bit of a given data block, $n = N$, over 19 blocks, 6 bits equally distributed in each block, in a diagonal way over consecutive blocks.

Or in other words the interleaving is a distribution of the encoded, reordered 456 bits from four given input data blocks, which taken together give $n=N$, over 22 bursts, 6 bits equally distributed in the first and 22nd bursts, 12 bits distributed in the second and 21st bursts, 18 bits distributed in the third and 20th bursts and 24 bits distributed in the other 16 bursts.

The block of coded data is interleaved "diagonal", where a new block of coded data starts with every fourth burst and is distributed over 22 bursts.

3.3.5 Mapping on a Burst

The mapping is done as specified for TCH/FS in section 3.1.4. On bitstealing by a FACCH, see section 4.2.5.

3.4 Data channel at full rate, 6.0 kbit/s radio interface rate (4.8 kbit/s services (TCH/F4.8))

The definition of a 6.0 kbit/s radio interface rate data flow for data services is given in GSM 04.21.

3.4.1 Interface with user unit

The user unit delivers to the encoder a bit stream organized in blocks of 60 information bits (data frames) every 10 ms, $\{d(0), \dots, d(59)\}$.

In the case where the user unit delivers to the encoder a bit stream organised in blocks of 240 information bits every 40 ms (e.g. RLP frames), the bits $\{d(0), \dots, d(59), d(60), \dots, d(60+59), d(2*60), \dots, d(2*60+59), d(3*60), \dots, d(3*60+59)\}$ shall be treated as four blocks of 60 bits each as described in the remainder of this section. To ensure end-to-end synchronisation of the 240 bit blocks, the resulting block after coding of the first 120 bits $\{d(0), \dots, d(60+59)\}$ shall be transmitted in one of the transmission blocks B0, B2, B4 of the channel mapping defined in GSM 05.02.

3.4.2 Block code

Sixteen bits equal to 0 are added to the 60 information bits, the result being a block of 76 bits, $\{u(0), \dots, u(75)\}$, with:

$$\begin{aligned} u(19k+p) &= d(15k+p) && \text{for } k = 0, \dots, 3 \text{ and } p = 0, \dots, 14; \\ u(19k+p) &= 0 && \text{for } k = 0, \dots, 3 \text{ and } p = 15, \dots, 18. \end{aligned}$$

Two such blocks forming a block of 152 bits $\{u'(0), \dots, u'(151)\}$ are dealt with together in the rest of the coding process

$$\begin{aligned} u'(k) &= u_1(k), && k=0, \dots, 75 \text{ (} u_1 = 1\text{:st block)} \\ u'(k+76) &= u_2(k), && k=0, \dots, 75 \text{ (} u_2 = 2\text{:nd block)} \end{aligned}$$

3.4.3 Convolutional encoder

This block of 152 bits is encoded with the convolutional code of rate 1/3 defined by the following polynomials:

$$\begin{aligned} G_1 &= 1 + D + D^3 + D^4 \\ G_2 &= 1 + D^2 + D^4 \\ G_3 &= 1 + D + D^2 + D^3 + D^4 \end{aligned}$$

The result is a block of $3 * 152 = 456$ coded bits, $\{c(0), \dots, c(455)\}$,

$$\begin{aligned} c(3k) &= u'(k) + u'(k-1) + u'(k-3) + u'(k-4) \\ c(3k+1) &= u'(k) + u'(k-2) + u'(k-4) \\ c(3k+2) &= u'(k) + u'(k-1) + u'(k-2) + u'(k-3) + u'(k-4) \end{aligned} \quad \text{for } k = 0, \dots, 151 ; u'(k) = 0, k < 0$$

3.4.4 Interleaving

The interleaving is done as specified for the TCH/F.9.6 in section 3.3.4

3.4.5 Mapping on a Burst

The mapping is done as specified for the TCH/FS in section 3.1.4. On bitstealing for signalling purposes by a FACCH, see section 4.2.5.

3.5 Data channel at half rate, 6.0 kbit/s radio interface rate (4.8 kbit/s services (TCH/H4.8))

The definition of a 6.0 kbit/s radio interface rate data flow for data services is given in GSM 04.21.

3.5.1 Interface with user unit

The user unit delivers to the encoder a bit stream organized in blocks of 60 information bits (data frames) every 10 ms. Four such blocks are dealt with together in the coding process $\{d(0), \dots, d(239)\}$.

For non transparent services those four blocks shall align with one complete 240-bit RLP frame.

3.5.2 Block code

The block encoding is done as specified for the TCH/F9.6 in section 3.3.2.

3.5.3 Convolutional encoder

The convolutional encoding is done as specified for the TCH/F9.6 in section 3.3.3.

3.5.4 Interleaving

The interleaving is done as specified for the TCH/F9.6 in section 3.3.4.

3.5.5 Mapping on a Burst

The mapping is done as specified for the TCH/FS in section 3.1.4. On bitstealing for signalling purposes by a FACCH, see section 4.3.5.

3.6 Data channel at full rate, 3.6 kbit/s radio interface rate (2.4 kbit/s and less services (TCH/F2.4))

The definition of a 3.6 kbit/s radio interface rate data flow for data services is given in GSM 04.21.

3.6.1 Interface with user unit

The user unit delivers to the encoder a bit stream organized in blocks of 36 information bits (data frames) every 10 ms. Two such blocks are dealt with together in the coding process $\{d(0), \dots, d(71)\}$.

3.6.2 Block code

This block of 72 information bits is not encoded, but only increased with four tail bits equal to 0 at the end of the block.

$$u(k) = d(k), k = 0, \dots, 71$$

$$u(k) = 0, k = 72, \dots, 75$$

3.6.3 Convolutional encoder

This block of 76 bits $\{u(0), \dots, u(75)\}$ is encoded with the convolutional code of rate 1/6 defined by the following polynomials:

$$G1 = 1 + D + D^3 + D^4$$

$$G2 = 1 + D^2 + D^4$$

$$G3 = 1 + D + D^2 + D^3 + D^4$$

$$G1 = 1 + D + D^3 + D^4$$

$$G2 = 1 + D^2 + D^4$$

$$G3 = 1 + D + D^2 + D^3 + D^4$$

The result is a block of 456 coded bits:

$\{c(0), c(1), \dots, c(455)\}$, defined by

$$\begin{aligned} c(6k) &= c(6k+3) = u(k) + u(k-1) + u(k-3) + u(k-4) \\ c(6k+1) &= c(6k+4) = u(k) + u(k-2) + u(k-4) \\ c(6k+2) &= c(6k+5) = u(k) + u(k-1) + u(k-2) + u(k-3) + u(k-4), \end{aligned} \quad \begin{array}{l} \text{for } k = 0, \dots, 75; \\ u(k) = 0, k < 0 \end{array}$$

3.6.4 Interleaving

The interleaving is done as specified for the TCH/FS in section 3.1.3

3.6.5 Mapping on a Burst

The mapping is done as specified for the TCH/FS in section 3.1.4.

3.7 Data channel at half rate, 3.6 kbit/s radio interface rate (2.4 kbit/s and less services (TCH/H2.4))

The definition of a 3.6 kbit/s radio interface rate data flow for data services is given in GSM 04.21

3.7.1 Interface with user unit

The user unit delivers to the encoder a bit stream organized in blocks of 36 information bits (data frames) every 10 ms. Two such blocks are dealt with together in the coding process, $\{d(0), \dots, d(71)\}$.

3.7.2 Block code

The block of 72 information bits is not encoded, but only increased with 4 tail bits equal to 0, at the end of the block.

Two such blocks forming a block of 152 bits $\{u(0), \dots, u(151)\}$ are dealt with together in the rest of the coding process.

$$\begin{array}{lll} u(k) & = d1(k), & k = 0, \dots, 75 \text{ (} d1 = \text{1st information block)} \\ u(k+76) & = d2(k), & k = 0, \dots, 75 \text{ (} d2 = \text{2nd information block)} \\ u(k) & = 0, & k = 72, \dots, 75, 148, \dots, 151 \end{array}$$

3.7.3 Convolutional encoder

The convolutional encoding is done as specified for the TCH/F4.8 in section 3.4.3.

3.7.4 Interleaving

The interleaving is done as specified for the TCH/F9.6 in section 3.3.4.

3.7.5 Mapping on a Burst

The mapping is done as specified for the TCH/FS in section 3.1.4. On bit stealing for signalling purposes by a FACCH, see section 4.3.5.

4. Control Channels

4.1 Slow associated control channel (SACCH)

4.1.1 Block constitution

The message delivered to the encoder has a fixed size of 184 information bits $\{d(0), \dots, d(183)\}$. It is delivered on a burst mode.

4.1.2 Block code

The block of 184 information bits is encoded, using a shortened binary cyclic code (FIRE code), with the following generator polynomial :

$$g(D) = (D^{23} + 1)(D^{17} + D^3 + 1)$$

The encoding of the cyclic code is performed in a systematic form, which means that, in GF(2), the polynomial:

$$u(0)D^{223} + u(1)D^{222} + \dots + u(222)D + u(223)$$

where $\{u(0), u(1), \dots, u(183)\}$ are the information bits ($u(k)=d(k)$, $k=0, \dots, 183$) and $\{u(184), u(185), \dots, u(223)\}$ are the parity bits ($u(k+184)=p(k)$, $k=0, \dots, 39$), when divided by $g(D)$ yields a remainder equal to $1 + D + D^2 + \dots + D^{39}$.

The result is a block of 224 bits, completed by 4 tail bits equal to 0 at the end of the block : $\{u(0), u(1), \dots, u(223)\}$, with $u(224)$ to $u(227)$ equal to 0.

4.1.3 Convolutional encoder

This block of 228 bits is encoded with the 1/2 rate convolutional code (identical to the one used for TCH/FS) defined by the polynomials:

$$\begin{aligned} G_0 &= 1 + D^3 + D^4 \\ G_1 &= 1 + D + D^3 + D^4 \end{aligned}$$

This results in a block of 456 coded bits: $\{c(0), \dots, c(455)\}$ defined by

$$\begin{aligned} c(2k) &= u(k) + u(k-3) + u(k-4) \\ c(2k+1) &= u(k) + u(k-1) + u(k-3) + u(k-4) \quad \text{for } k = 0, \dots, 227 ; u(k) = 0, k < 0 \end{aligned}$$

4.1.4 Interleaving

The coded bits are reordered and interleaved according to the following rule

$$\begin{aligned} i(B, j) &= c(n, k) \quad \text{for} \quad \begin{aligned} k &= 0, 1, \dots, 455 \\ n &= 0, 1, \dots, N, N+1, \dots \\ B &= B_0 + 4n + k \text{ mod } (4) \\ j &= 2[(49k) \text{ mod } 57] + [(k \text{ mod } 8) \text{ div } 4] \end{aligned} \end{aligned}$$

The result of the reordering of bits is the same as given for a TCH/FS (section 3.1.3) as can be seen from the evaluation of the bit number-index j , distributing the 456 bits over 4 blocks on even numbered bits and 4 blocks on odd numbered bits. The resulting 4 blocks are built by putting blocks with even numbered bits and blocks with odd numbered bits together into one block.

The block of coded data is interleaved "block rectangular" where a new data block starts every 4th block and is distributed over 4 blocks.

4.1.5 Mapping on a Burst

The mapping is given by the rule

$$\begin{aligned} e(B, j) &= i(B, j) \quad \text{and} \quad e(B, 59+j) = i(B, 57+j) \quad \text{for } j = 0, \dots, 56 \\ \text{and} \quad e(B, 57) &= hl(B) \quad \text{and} \quad e(B, 58) = hu(B) \end{aligned}$$

The two bits labelled $hl(B)$ and $hu(B)$ on burst number B are flags used for indication of control channel signalling. They are set to "1" for a SACCH.

4.2 Fast associated control channel at full rate (FACCH/F)

4.2.1 Block constitution

The message delivered to the encoder has a fixed size of 184 information bits. It is delivered on a burst mode.

4.2.2 Block code

The block encoding is done as specified for the SACCH in section 4.1.2.

4.2.3 Convolutional encoder

The convolutional encoding is done as specified for the SACCH in section 4.1.3.

4.2.4 Interleaving

The interleaving is done as specified for the TCH/FS in section 3.1.3.

4.2.5 Mapping on a Burst

A FACCH/F frame of 456 coded bits is mapped on 8 consecutive bursts as specified for the TCH/FS in section 3.1.4. As a FACCH is transmitted on bits which are stolen in a burst from the traffic channel, the even numbered bits in the first 4 bursts and the odd numbered bits of the last 4 bursts are stolen.

To indicate this to the receiving device the flags $hl(B)$ and $hu(B)$ have to be set according to the following rule :

$hu(B) = 1$ for the first 4 burst (even numbered bits are stolen)
 $hl(B) = 1$ for the last 4 bursts (odd numbered bits are stolen)

The consequences of this bitstealing by a FACCH/F is for a

- speech channel (TCH/FS) and data channel (TCH/F2.4):
One full frame of data is stolen by the FACCH.
- Data channel (TCH/F9.6):
The bitstealing by a FACCH/F disturbs a maximum of 96 coded bits generated from an input frame of four data blocks. A maximum of 24 of the 114 coded bits resulting from one input data block of 60 bits may be disturbed.
- Data channel (TCH/F4.8):
The bit stealing by FACCH/F disturbs a maximum of 96 coded bits generated from an input frame of two data blocks. A maximum of 48 of the 228 coded bits resulting from one input data block of 60 bits may be disturbed.

NOTE: In the case of consecutive stolen frames, a number of bursts will have both the even and the odd bits stolen and both flags $hu(B)$ and $hl(B)$ must be set to 1.

4.3 Fast associated control channel at half rate (FACCH/H)

4.3.1 Block constitution

The message delivered to the encoder has a fixed size of 184 information bits. It is delivered on a burst mode.

4.3.2 Block code

The block encoding is done as specified for the SACCH in section 4.1.2.

4.3.3 Convolutional encoder

The convolutional encoding is done as specified for the SACCH in section 4.1.3.

4.3.4 Interleaving

The coded bits are reordered and interleaved according to the following rule:

$$i(B,j) = c(n,k), \quad \text{for } k=0,1,\dots,455$$

$$n=0,1,\dots,N,N+1,\dots$$

$$B = B_0 + 4n + k \bmod (8) - 4[(k \bmod 8) \text{ div } 6]$$

$$j = 2[(49k) \bmod 57] + [(k \bmod 8) \text{ div } 4]$$

The result of the reordering of bits is the same as given for a TCH/FS (section 3.1.3) as can be seen from the evaluation of the bit number-index j , distributing the 456 bits over 4 blocks on even numbered bits and 4 blocks on odd numbered bits. The 2 last blocks with even numbered bits and the 2 last blocks with odd numbered bits are put together into 2 full middle blocks.

The block of coded data is interleaved "block diagonal" where a new data block starts every 4th block and is distributed over 6 blocks.

4.3.5 Mapping on a Burst

A FACCH/H frame of 456 coded bits is mapped on 6 consecutive bursts by the rule:

$$e(B,j)=i(B,j) \quad \text{and} \quad e(B,59+j)=i(B,57+j) \quad \text{for } j=0,\dots,56$$

and

$$e(B,57)=hl(B) \quad \text{and} \quad e(B,58)=hu(B)$$

As a FACCH/H is transmitted on bits which are stolen from the traffic channel, the even numbered bits of the first 2 bursts, all bits of the middle 2 bursts and the odd numbered bits of the last 2 bursts are stolen.

To indicate this to the receiving device the flags $hl(B)$ and $hu(B)$ have to be set according to the following rule:

$$hu(B) = 1 \quad \text{for the first 2 bursts (even numbered bits are stolen)}$$

$$hu(B) = 1 \quad \text{and} \quad hl(B) = 1 \quad \text{for the middle 2 bursts (all bits are stolen)}$$

$$hl(B) = 1 \quad \text{for the last 2 bursts (odd numbered bits are stolen)}$$

The consequences of this bitstealing by a FACCH/H is for a

- speech channel (TCH/HS):
Two full consecutive speech frames are stolen by a FACCH/H.
- data channel (TCH/H4.8):

The bitstealing by FACCH/H disturbs a maximum of 96 coded bits generated from an input frame of four data blocks. A maximum of 24 out of the 114 coded bits resulting from one input data block of 60 bits may be disturbed.

- data channel (TCH/H2.4):

The bitstealing by FACCH/H disturbs a maximum of 96 coded bits generated from an input frame of four data blocks. A maximum of 24 out of the 114 coded bits resulting from one input data block of 36 bits may be disturbed.

NOTE: In the case of consecutive stolen frames, two overlapping bursts will have both the even and the odd numbered bits stolen and both flags hu(B) and hl(B) must be set to 1.

4.4 Broadcast, Paging, Access grant and Cell broadcast channels (BCCH, PCH, AGCH, CBCH)

The coding scheme used for the broadcast, paging, access grant and cell broadcast messages is the same as for the SACCH messages, specified in 4.1.

4.5 Stand-alone dedicated control channel (SDCCH)

The coding scheme used for the dedicated control channel messages is the same as for SACCH messages, specified in 4.1.

4.6 Random access channel (RACH)

The burst carrying the random access uplink message has a different structure. It contains 8 information bits $d(0), \dots, d(7)$.

Six parity bits $p(0)$ to $p(5)$ are defined in such a way that in $GF(2)$ the binary polynomial $d(0)D^{13} + \dots + d(7)D^6 + p(0)D^5 + \dots + p(5)$, when divided by $D^6 + D^5 + D^3 + D^2 + D + 1$ yields a remainder equal to $D^5 + D^4 + D^3 + D^2 + D + 1$.

The six bits of the BSIC, $B(0)$ to $B(5)$, of the BS to which the Random Access is intended, are added bitwise modulo 2 to the six parity bits. $P(0)$ to $P(5)$. This results in six colour bits, $C(0)$ to $C(5)$ defined as $C(k) = b(k) + p(k)$ ($k = 0$ to 5) where

$b(0)$ = MSB of PLMN colour code
 $b(5)$ = LSB of BS colour code.

This defines $\{u(0), \dots, u(17)\}$ by:

$u(k) = d(k)$ ($k = 0$ to 7)
 $u(8+k) = c(k)$ ($k = 0$ to 5)
 $u(14+k) = 0$ ($k = 0$ to 3) (tail bits)

The bits $\{e(0), \dots, e(35)\}$ are obtained by the same convolutional code of rate 1/2 as for TCH/FS, defined by the polynomials:

$G_0 = 1 + D^3 + D^4$
 $G_1 = 1 + D + D^3 + D^4$

and with

$e(2k) = u(k) + u(k-3) + u(k-4)$
 $e(2k+1) = u(k) + u(k-1) + u(k-3) + u(k-4)$ for $k = 0, \dots, 17$; $u(k) = 0$, $k < 0$

4.7 Synchronization channel (SCH)

The burst carrying the synchronization information on the downlink BCCH has a different structure. It contains 25 information bits $\{d(0), \dots, d(24)\}$, 10 parity bits $\{p(0), \dots, p(9)\}$ and 4 tail bits. The precise ordering of the information bits is given in GSM 04.08.

The ten parity bits $p(0)$ to $p(9)$ are defined in such a way that in $GF(2)$ the binary polynomial:

$d(0)D^{34} + \dots + d(24)D^{10} + p(0)D^9 + \dots + p(9)$, when divided by:

$D^{10} + D^8 + D^6 + D^5 + D^4 + D^2 + 1$, yields a remainder equal to:

$$D^9 + D^8 + D^7 + D^6 + D^5 + D^4 + D^3 + D^2 + D + 1.$$

Thus the encoded bits $\{u(0), \dots, u(38)\}$ are:

$$\begin{aligned} u(k) &= d(k) \quad (k = 0 \text{ to } 24) \\ u(25+k) &= p(k) \quad (k = 0 \text{ to } 9) \\ u(35+k) &= 0 \quad (k = 0 \text{ to } 3) \text{ (tail bits)} \end{aligned}$$

The bits $\{e(0), \dots, e(77)\}$ are obtained by the same convolutional code of rate 1/2 as for TCH/FS, defined by the polynomials:

$$\begin{aligned} G_0 &= 1 + D^3 + D^4 \\ G_1 &= 1 + D + D^3 + D^4 \end{aligned}$$

and with

$$\begin{aligned} e(2k) &= u(k) + u(k-3) + u(k-4) \\ e(2k+1) &= u(k) + u(k-1) + u(k-3) + u(k-4) \text{ for } k = 0, \dots, 77; u(k) = 0, k < 0 \end{aligned}$$

4.8 Handover Access Burst

The encoding of this burst is as defined in section 4.6 for the random access channel (RACH). The BSIC used shall be the BSIC of the BS to which the HO is done.

Table 1: Reordering and Partitioning a Coded Block into 8 Subblocks

j =		0	1	2	3	4	5	6	7
i =	0	k = 0	57	114	171	228	285	342	399
		64	121	178	235	292	349	406	7
		128	185	242	299	356	413	14	71
		192	249	306	363	420	21	78	135
		256	313	370	427	28	85	142	199
	5	320	377	434	35	92	149	206	263
		384	441	42	99	156	213	270	327
		448	49	106	163	220	277	334	391
		56	113	170	227	284	341	398	455
	10	120	177	234	291	348	405	6	63
		184	241	298	355	412	13	70	127
		248	305	362	419	20	77	134	191
		312	369	426	27	84	141	198	255
		376	433	34	91	148	205	262	319
	15	440	41	98	155	212	269	326	383
		48	105	162	219	276	333	390	447
		112	169	226	283	340	397	454	55
		176	233	290	347	404	5	62	119
		240	297	354	411	12	69	126	183
	20	304	361	418	19	76	133	190	247
		368	425	26	83	140	197	254	311
		432	33	90	147	204	261	318	375
		40	97	154	211	268	325	382	439
		104	161	218	275	332	389	446	47
	25	168	225	282	339	396	453	54	111
		232	289	346	403	4	61	118	175
		296	353	410	11	68	125	182	239
		360	417	18	75	132	189	246	303
		424	25	82	139	196	253	310	367
	30	32	89	146	203	260	317	374	431
		96	153	210	267	324	381	438	39
		160	217	274	331	388	445	46	103
		224	281	338	395	452	53	110	167
		288	345	402	3	60	117	174	231
	35	352	409	10	67	124	181	238	295
		416	17	74	131	188	245	302	359
		24	81	138	195	252	309	366	423
		88	145	202	259	316	373	430	31
		152	209	266	323	380	437	38	95
	40	216	273	330	387	444	45	102	159
		280	337	394	451	52	109	166	223
		344	401	2	59	116	173	230	287
		408	9	66	123	180	237	294	351
		16	73	130	187	244	301	358	415
	45	80	137	194	251	308	365	422	23
		144	201	258	315	372	429	30	87
		208	265	322	379	436	37	94	151
		272	329	386	443	44	101	158	215
		336	393	450	51	108	165	222	279
	50	400	1	58	115	172	229	286	343
		8	65	122	179	236	293	350	407
		72	129	186	243	300	357	414	15
		136	193	250	307	364	421	22	79
		200	257	314	371	428	29	86	143
		264	321	378	435	36	93	150	207
	55	328	385	442	43	100	157	214	271
	56	392	449	50	107	164	221	278	335

Table 2: subjective importance of encoded bits (Parameter names and bit indices refer to 06.10)

Importance class	Parameter name	Parameter number	Bit index	Label	Class
1	Log area ratio 1 block amplitude	1	5	d0 d1, d2, d3, d4	1
		12, 29, 46, 63	5		
2	Log area ratio 1 Log area ratio 2 Log area ratio 3	1	4	d0 d1, d2, d3, d4 ...d48, d49 d50... ...d181 d182	
		2	5		
		3	4		
3	Log area ratio 1 Log area ratio 2 Log area ratio 3 Log area ratio 4 LTP lag Block amplitude Log area ratio 2,5,6 LTP lag LTP lag LTP lag LTP lag	1	3		
		2	4		
		3	3		
		4	4		
		9, 26, 43,60	6		
		12, 29, 46, 63	4		
		2, 5, 6	3		
		9, 26, 43, 60	5		
		9, 26, 43, 60	4		
		9, 26, 43, 60	3		
9, 26, 43, 60	2				
4	Block amplitude Log area ratio 1 Log area ratio 4 Log area ratio 7 LTP lag Log area ratio 5,6 LTP gain LTP lag Grid position	12, 29, 46, 63	3		
		1	2		
		4	3		
		7	2		
		9, 26, 43, 60	1		
		5, 6	2		
		10, 27, 44, 61	1		
		9, 26, 43, 60	0		
11, 28, 45, 62	1				
5	Log area ratio 1 Log area ratio 2, 3, 8, 4 Log area ratio 5, 7 LTP gain Block amplitude RPE pulses RPE pulses RPE pulses RPE pulses Grid position Block amplitude RPE pulses RPE pulses RPE pulses RPE pulses RPE pulses	1	1		
		2, 3, 8, 4	2		
		5, 7	1		
		10, 27, 44, 61	0		
		12, 29, 46, 63	2		
		13..25	2		
		30..42	2		
		47..59	2		
		64..76	2		
		11, 28, 45, 62	0		
		12, 29, 46, 63	1		
		13..25	1		
		30..42	1		
		47..59	1		
		64..67	1		
68..76	1				
6	Log area ratio 1 Log area ratio 2, 3, 5 Log area ratio 7 Log area ratio 8 Log area ratio 8, 3 Log area ratio 4 Log area ratio 4, 5 Block amplitude RPE pulses RPE pulses RPE pulses RPE pulses Log area ratio 2, 5	1	0		
		2, 3, 5	1		
		7	0		
		8	1		
		8, 3	0		
		4	1		
		4, 5	0		
		12, 29, 46, 63	0		
		13..25	0		
		30..42	0		
		47..59	0		
		64..76	0		
		2,5	0		

Annex A (informative): Summary of Channel Types

TCH/FS:	full rate speech traffic channel
TCH/F9.6:	9.6 kbit/s full rate data traffic channel
TCH/F4.8:	4.8 kbit/s full rate data traffic channel
TCH/H4.8:	4.8 kbit/s half rate data traffic channel
TCH/F2.4:	≤ 2.4 kbit/s full rate data traffic channel
TCH/H2.4:	≤ 2.4 kbit/s half rate data traffic channel
SACCH:	slow associated control channel
FACCH/F:	fast associated control channel at full rate
FACCH/H:	fast associated control channel at half rate
SDCCH:	stand-alone dedicated control channel
BCCH:	broadcast control channel
PCH:	paging channel
AGCH:	access grant channel
RACH:	random access channel
SCH:	synchronization channel
CBCH:	cell broadcast channel

Annex B (informative): Summary of Polynomials Used for Convolutional Codes

$G_0 = 1 + D^3 + D^4$	TCH/FS, TCH/F9.6, TCH/H4.8, SDCCH, BCCH, PCH, SACCH, FACCH, AGCH, RACH, SCH
$G_1 = 1 + D + D^3 + D^4$	TCH/FS, TCH/F9.6, TCH/H4.8, SACCH, FACCH, SDCCH, BCCH, PCH, AGCH, RACH, SCH, TCH/F4.8, TCH/F2.4, TCH/H2.4
$G_2 = 1 + D^2 + D^4$	TCH/F4.8, TCH/F2.4, TCH/H2.4
$G_3 = 1 + D + D^2 + D^3 + D^4$	TCH/F4.8, TCH/F2.4, TCH/H2.4

History

Document history	
September 1994	First Edition
April 1996	Converted into Adobe Acrobat Portable Document Format (PDF)